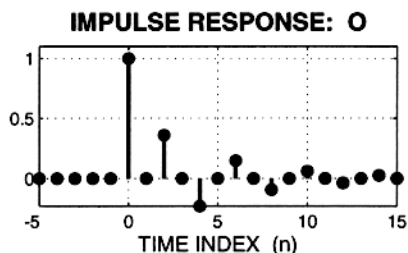
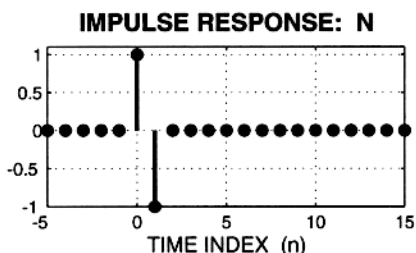
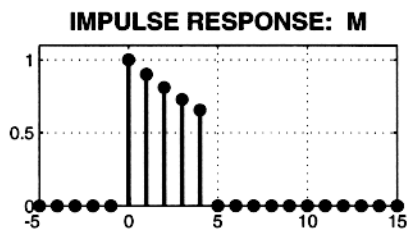
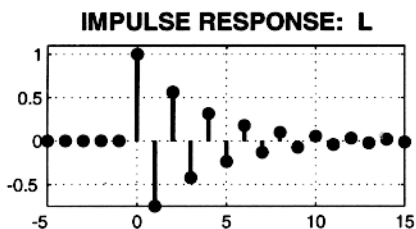
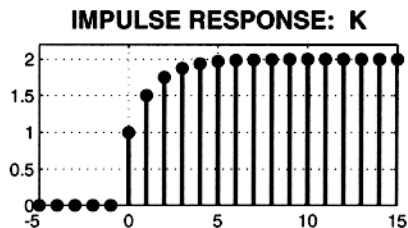
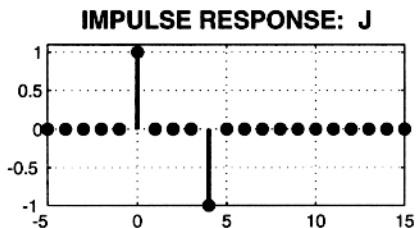


Problem fall-99-F.1:



For each of the impulse-response plots (J, K, L, M, N, O), determine which one of the following systems (specified by either an $H(z)$ or a difference equation) matches the impulse response.

$$S_1: y[n] = -0.75y[n-1] + x[n]$$

$$S_2: H(z) = \frac{1 + z^{-2}}{1 + 0.64z^{-2}}$$

$$S_3: H(z) = \sum_{k=0}^4 z^{-k}$$

$$S_4: H(z) = \frac{1 + z^{-2}}{1 - 0.75z^{-1}}$$

$$S_5: H(z) = \frac{2}{1 - z^{-1}} + \frac{-1}{1 - 0.5z^{-1}}$$

$$S_6: H(z) = \sum_{k=0}^4 (0.9)^k z^{-k}$$

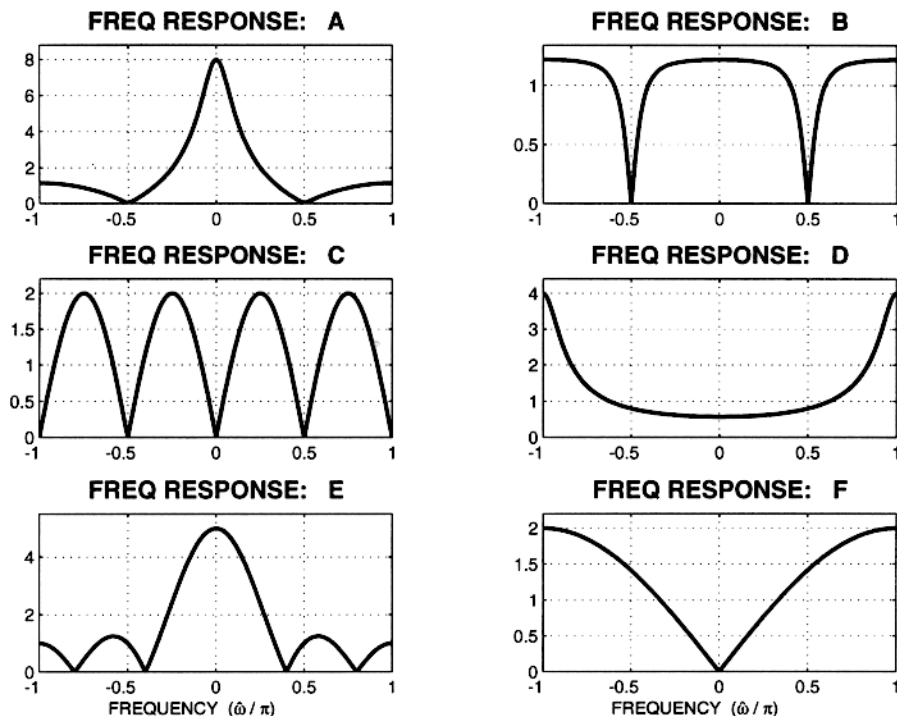
$$S_7: y[n] = x[n] - x[n-1]$$

$$S_8: H(z) = 1 - z^{-4}$$

Mark your answer in the following table:

| IMPULSE RESPONSE | SYSTEM ($S_{\#}$) | IMPULSE RESPONSE | SYSTEM ($S_{\#}$) |
|------------------|---------------------|------------------|---------------------|
| J | 8 | K | 5 |
| L | 1 | M | 6 |
| N | 7 | O | 2 |

Problem fall-99-F.2:



For each of the frequency response plots (A, B, C, D, E, F), determine which one of the following systems (specified by either an $H(z)$ or a difference equation) matches the frequency response (magnitude only). NOTE: frequency axis is **normalized**; it is $\hat{\omega}/\pi$.

$$S_1: y[n] = -0.75y[n-1] + x[n]$$

$$S_2: H(z) = \frac{1+z^{-2}}{1+0.64z^{-2}}$$

$$S_3: H(z) = \sum_{k=0}^4 z^{-k}$$

$$S_4: H(z) = \frac{1+z^{-2}}{1-0.75z^{-1}}$$

$$S_5: H(z) = \frac{2}{1-z^{-1}} + \frac{-1}{1-0.5z^{-1}}$$

$$S_6: H(z) = \sum_{k=0}^4 (0.9)^k z^{-k}$$

$$S_7: y[n] = x[n] - x[n-1]$$

$$S_8: H(z) = 1 - z^{-4}$$

Mark your answer in the following table:

| FREQUENCY RESPONSE | SYSTEM ($S_{\#}$) | FREQUENCY RESPONSE | SYSTEM ($S_{\#}$) |
|--------------------|---------------------|--------------------|---------------------|
| A | 4 | B | 2 |
| C | 8 | D | 1 |
| E | 3 | F | 7 |

Problem fall-99-F.3:

A discrete-time system is defined by the following system function:

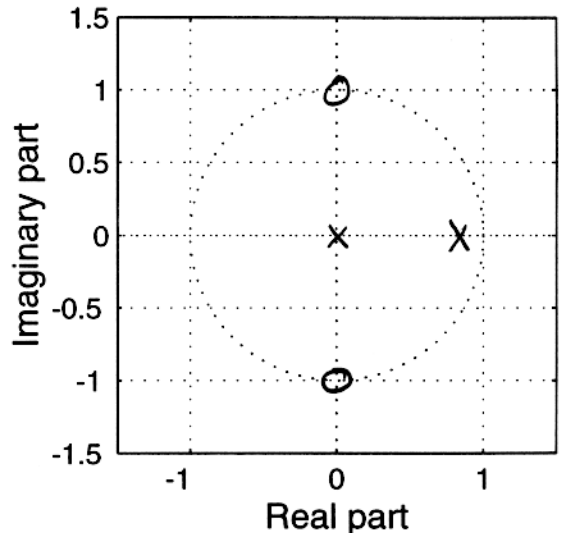
$$H(z) = \frac{1 + z^{-2}}{1 - 0.75z^{-1}} = \frac{1}{1 - 0.75z^{-1}} + \frac{z^{-2}}{1 - 0.75z^{-1}}$$

- (a) Use the first form of $H(z)$ to determine *all* the poles and zeros of $H(z)$ and plot them in the z -plane.

$$\begin{aligned} H(z) &= \frac{z^2 + 1}{z(z - 0.75)} \\ &= \frac{(z - j)(z + j)}{z(z - 0.75)} \end{aligned}$$

Zeros at $z = \pm j$

Poles at $z = 0, .75$



- (b) Use the second form of $H(z)$ above to find the corresponding impulse response $h[n]$.

$$h[n] = (.75)^n u[n] + (.75)^{n-2} u[n-2]$$

- (c) Use the first form of $H(z)$ to obtain an expression for the magnitude-squared of the frequency response $|H(e^{j\hat{\omega}})|^2 = H(e^{j\hat{\omega}})H^*(e^{j\hat{\omega}})$. Your answer should involve only real quantities.

$$\begin{aligned} |H(e^{j\hat{\omega}})|^2 &= \left(\frac{1 + e^{-j2\hat{\omega}}}{1 - .75e^{-j\hat{\omega}}} \right) \left(\frac{1 + e^{j2\hat{\omega}}}{1 - .75e^{j\hat{\omega}}} \right) \\ &= \frac{1 + 2\cos(2\hat{\omega}) + 1}{1 - 1.5\cos\hat{\omega} + (.75)^2} = \frac{2(1 + \cos(2\hat{\omega}))}{1.5625 - 1.5\cos\hat{\omega}} \end{aligned}$$

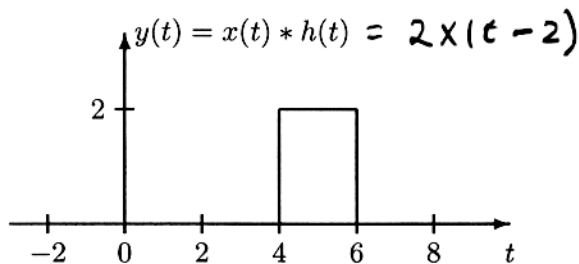
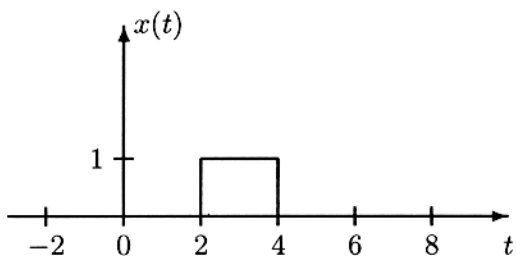
- (d) For what value of $\hat{\omega}$ will it be true that $y[n] = 0$ for $-\infty < n < \infty$ when the input to the system is $x[n] = e^{j\hat{\omega}n}$ for $-\infty < n < \infty$?

Since we have a zero at $z = \pm j = e^{j\pm\pi/2}$

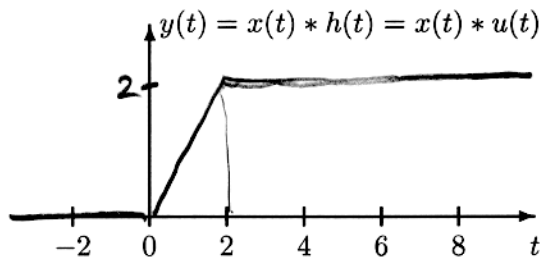
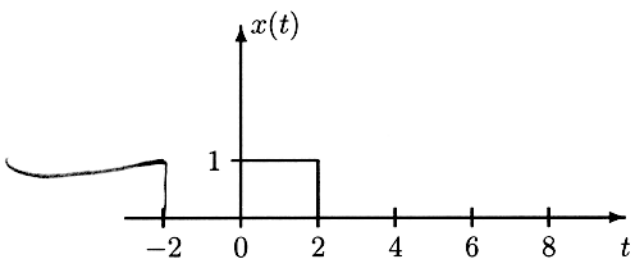
$$\hat{\omega} = \pm \pi/2$$

Problem fall-99-F.4:

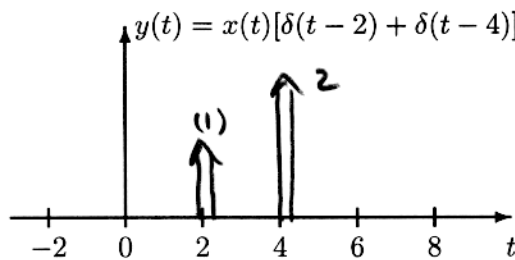
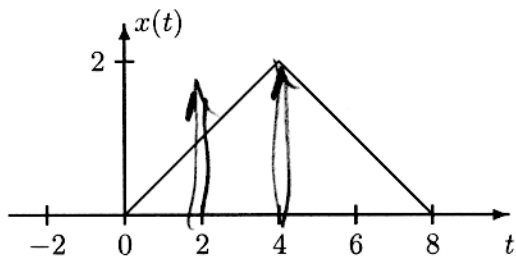
(a) Given that $y(t) = x(t) * h(t)$, find $h(t)$. $h(t) = \underline{2 \delta(t-2)}$



(b) If $h(t) = u(t)$, plot $y(t) = x(t) * h(t)$ on the graph on the right. Be sure to label the $y(t)$ axis.

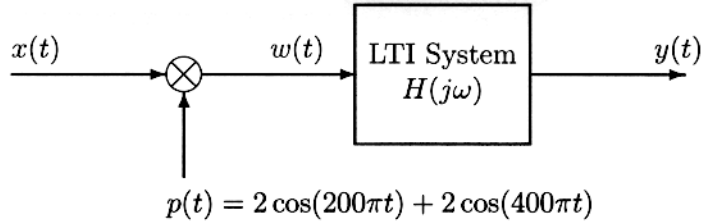


(c) Plot $y(t) = x(t)[\delta(t-2) + \delta(t-4)]$ on the graph on the right.

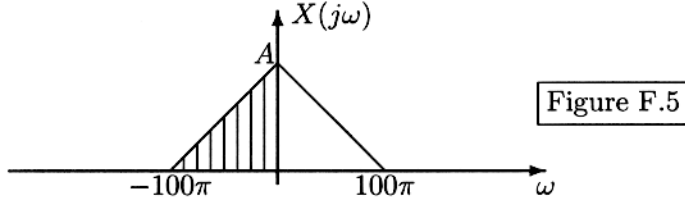


$$\begin{aligned}
 y(t) &= x(t) [\delta(t-2) + \delta(t-4)] \\
 &= x(t) \delta(t-2) + x(t) \delta(t-4) \\
 &= x(2) \delta(t-2) + x(4) \delta(t-4) \\
 &= \delta(t-2) + 2 \delta(t-4)
 \end{aligned}$$

Problem fall-99-F.5:



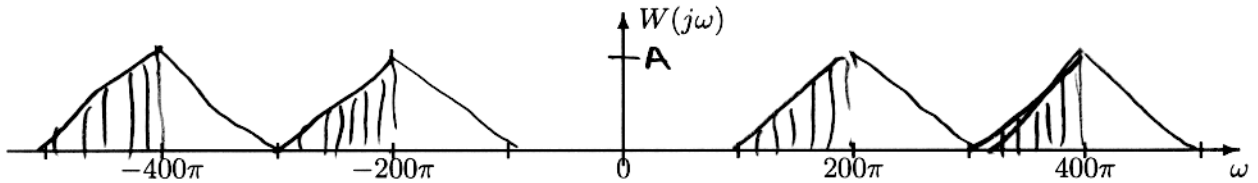
In the above modulation/filtering system, assume that the input signal $x(t)$ has a bandlimited Fourier transform $X(j\omega)$ as depicted in **Figure F.5** below.



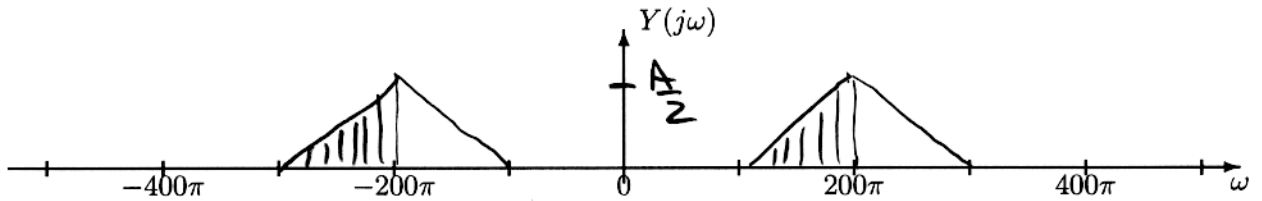
- (a) Give the general equation that expresses $W(j\omega)$, the Fourier transform of $w(t) = x(t)[2 \cos(200\pi t) + 2 \cos(400\pi t)]$, in terms of $X(j\omega)$.

$$W(j\omega) = X(j(\omega - 200\pi)) + X(j(\omega + 200\pi)) + X(j(\omega - 400\pi)) + X(j(\omega + 400\pi))$$

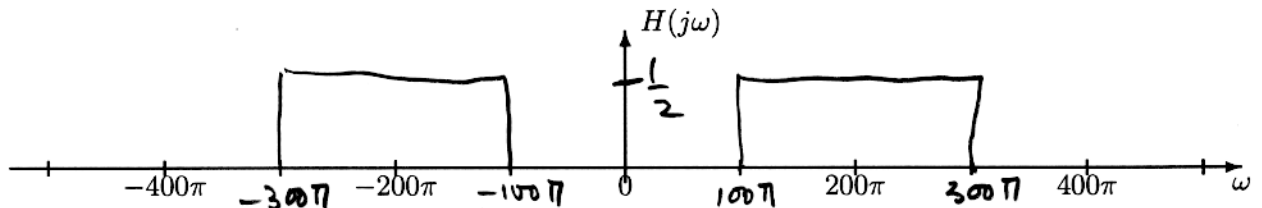
Also, plot the Fourier transform $W(j\omega)$ for the specific input $x(t)$ whose Fourier transform $X(j\omega)$ is given above in **Figure F.5**. Note that the negative frequency portion of the Fourier transform $X(j\omega)$ is shaded. Mark the corresponding region or regions in your plot of $W(j\omega)$, and be sure to carefully label both amplitudes and frequencies.



- (b) It is desired that the output of the filter be $y(t) = x(t) \cos(200\pi t)$. Plot the Fourier transform $Y(j\omega)$ below for the $X(j\omega)$ given in **Figure F.5** above. Be sure to carefully label both amplitudes and frequencies.



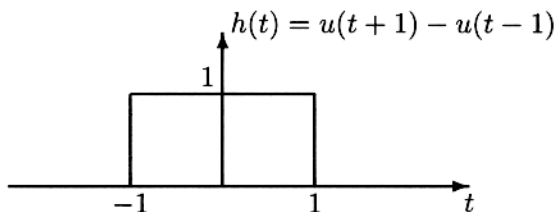
- (c) Plot the frequency response $H(j\omega)$ of a filter that is required to obtain $y(t)$ from $w(t)$. Be sure to carefully label the cutoff frequency(s) and gain of the filter.



Lowpass also works.

Problem fall-99-F.6:

The impulse response of an LTI system is

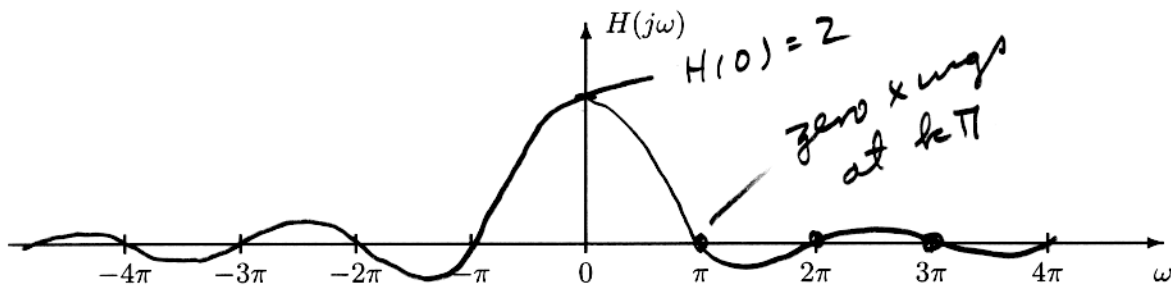


- (a) Determine the frequency response $H(j\omega)$ of the system.

$$H(j\omega) = \int_{-1}^1 e^{-j\omega t} dt = \left. \frac{e^{-j\omega t}}{-j\omega} \right|_{-1}^1 = \underline{\underline{2 \frac{\sin \omega}{\omega}}}$$

or look up on sheet

- (b) Make a carefully labeled sketch of $H(j\omega)$ on the axes below.



- (c) Is the LTI system stable? Explain your answer to receive full credit.

yes it is stable since

$$\int_{-1}^1 1 dt = 2 < \infty$$

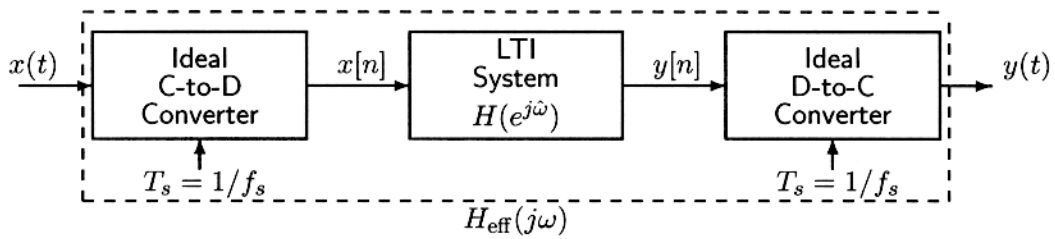
- (d) Is the LTI system causal? Explain your answer to receive full credit.

No it is not causal since

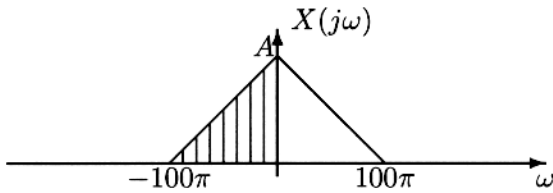
$$h(t) \neq 0 \quad t < 0$$

Problem fall-99-F.8:

Consider the following system for discrete-time filtering of a continuous-time signal:



- (a) In this part, assume that $y[n] = x[n]$ (i.e., the identity system) and assume that the input signal $x(t)$ has a bandlimited Fourier transform $X(j\omega)$ as depicted below. For this input signal, what is the *smallest* value of the sampling frequency f_s such that $y(t) = x(t)$?



smallest $f_s = 100$ samples/sec

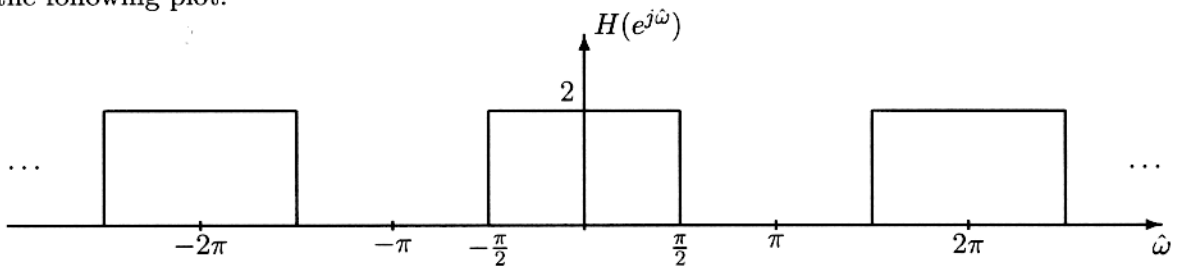
- (b) In this part again assume that $y[n] = x[n]$ (i.e., the identity system) but now assume that the input signal is $x(t) = 10 \cos(150\pi t + \pi/3)$. If the sampling rate is $f_s = 100$ samples/sec, what is the corresponding output $y(t)$?

$$X[n] = 10 \cos(150\pi n/100 + \pi/3) = 10 \cos(1.5\pi n + \pi/3)$$

$$= 10 \cos(2\pi n - 1.5\pi n + \pi/3) = 10 \cos(1.5\pi n - \pi/3)$$

$$\therefore y(t) = 10 \cos(1.5\pi \cdot 100t - \pi/3) = 10 \cos(50\pi t - \pi/3)$$

- (c) In this part, assume that the discrete-time system has frequency response $H(e^{j\hat{\omega}})$ defined by the following plot:



Now, if $f_s = 200$ samples/sec, make a carefully labeled plot below of $H_{\text{eff}}(j\omega)$, the effective frequency response of the overall system. Also plot $Y(j\omega)$, the Fourier transform of the output $y(t)$, when the input has Fourier transform $X(j\omega)$ as depicted in the graph of part (a).

$$H_{\text{eff}}(j\omega) = H(e^{j\omega/200})$$

