

ECE 7251: Signal Detection and Estimation

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Lecture 18, 2/18/02:
Wiener Filtering

The Setup

- Context: Bayesian linear MMSE estimation for random sequences
- Parameter sequence $\{\Theta_k, k \in \mathbb{Z}\}$
- Data sequence $\{Y_k, k \in I \subset \mathbb{Z}\}$

- Goal: Estimate $\{\mathbf{q}_k\}$ as a linear function of the observations:

$$\hat{\mathbf{q}}_k(y) = \sum_{j \in I} h(k, j) y_j$$

- Find h to minimize mean square error

O.P. to the Rescue

- By the orthogonality principle,

$$E[(\hat{\mathbf{q}}_k(Y) - \Theta_k) Y_i^*] = 0 \text{ for } i \in I$$

$$E[(\sum_{j \in I} h(k, j) Y_j - \Theta_k) Y_i^*] = 0$$

$$\sum_{j \in I} h(k, j) E[Y_j Y_i^*] = E[\Theta_k Y_i^*]$$

$$\sum_{j \in I} h(k, j) r_Y(j, i) = r_{\Theta Y}(k, i)$$

Wiener-Hopf Equation

- If processes are stationary, we can write

$$\sum_{j \in I} h(k, j) r_Y(j - i) = r_{\Theta Y}(k - i)$$

slight abuse of notation

Spectral Representation

- If $I = \mathbb{Z}$, it turns out the filter is LTI:

$$\sum_{j \in I} h(k - j) r_Y(j - i) = r_{\Theta Y}(k - i)$$

WLOG, consider $i=0$: $\sum_{j \in I} h(k - j) r_Y(j) = r_{\Theta Y}(k)$

- Can solve \mathbb{W} in the \mathbb{Z} transform domain:

$$H(z) S_Y(z) = S_{\Theta Y}(z)$$

$$H(z) = \frac{S_{\Theta Y}(z)}{S_Y(z)}$$

Mean Square Error

$$\begin{aligned} \bullet \text{MSE} &= E[(\Theta_k - \hat{\mathbf{q}}(Y))^2] \\ &= E[(\Theta_k - \hat{\mathbf{q}}(Y))(\Theta_k^* - \hat{\mathbf{q}}^*(Y))] \\ &= E[(\Theta_k - \hat{\mathbf{q}}(Y))\Theta_k^*] + E[(\Theta_k - \hat{\mathbf{q}}(Y))\hat{\mathbf{q}}^*(Y)] \quad (\text{by O.P.}) \\ &= E[\Theta_k \Theta_k^*] - E[\hat{\mathbf{q}}(Y)\Theta_k^*] \\ &= E[\Theta_k \Theta_k^*] - E[\sum_{j \in I} h(k - j) Y_j \Theta_k^*] \\ &= E[\Theta_k \Theta_k^*] - \sum_{j \in I} h(k - j) E[Y_j \Theta_k^*] \\ &= r_{\Theta}(0) - \sum_{j \in I} h(k - j) r_{Y\Theta}(j - k) \end{aligned}$$

Mean Square Error

$$\text{MSE} = r_{\Theta}(0) - \sum_{j \in I} h(k - j) r_{Y\Theta}(j - k)$$

- Since everything is stationary, can just take $k=0$

$$\text{MSE} = r_{\Theta}(0) - (h * r_{Y\Theta})(0)$$

$$= \int_{-p}^p S_{\Theta}(\mathbf{w}) - H(\mathbf{w}) S_{Y\Theta}(\mathbf{w}) d\mathbf{w}$$

$$= \int_{-p}^p S_{\Theta}(\mathbf{w}) - \frac{S_{\Theta Y}(\mathbf{w})}{S_Y(\mathbf{w})} S_{Y\Theta}(\mathbf{w}) d\mathbf{w}$$

$$= \int_{-p}^p S_{\Theta}(\mathbf{w}) - \frac{|S_{\Theta Y}(\mathbf{w})|^2}{S_Y(\mathbf{w})} d\mathbf{w}$$

Example: Deblurring

- Suppose object is observed through a blurring point spread function f and additive noise W

$$Y_k = (f * \Theta)_k + W_k$$

- Suppose Θ and W are uncorrelated zero-mean
- Recall from ECE6601:

$$S_Y = |F|^2 S_q + S_W \text{ and } S_{\Theta Y} = F^* S_Y$$

- So the Wiener filter is

$$H(z) = \frac{S_{\Theta Y}(z)}{S_Y(z)} = \frac{F^*(z)S_{\Theta}(z)}{|F(z)|^2 S_q(z) + S_W(z)}$$

Interpretation of Deblurring Filter

- If noise is negligible, i.e. $S_W(\mathbf{w}) \approx 0$

$$H(\mathbf{w}) = \frac{F^* S_q}{|F|^2 S_q + S_W} \approx \frac{F^* S_q}{F F^* S_q} = \frac{1}{F}$$

- Even if there is no noise, in implementation, straight division by $F(\omega)$ is often ill posed and not a good idea (round off errors, etc.)

Deblurring Error

$$\begin{aligned} MSE &= \int_{-p}^p S_{\Theta}(\mathbf{w}) - \frac{|S_{\Theta Y}(\mathbf{w})|^2}{S_Y(\mathbf{w})} d\mathbf{w} \\ &= \int_{-p}^p S_{\Theta}(\mathbf{w}) - \frac{|F(\mathbf{w})|^2 |S_{\Theta}(\mathbf{w})|^2}{|F(\mathbf{w})|^2 S_{\Theta}(\mathbf{w}) + S_W(\mathbf{w})} d\mathbf{w} \\ &= \int_{-p}^p \frac{S_{\Theta} [|F|^2 S_{\Theta} + S_W] - |F|^2 |S_{\Theta}|^2}{|F|^2 S_{\Theta} + S_W} d\mathbf{w} \\ &= \int_{-p}^p \frac{S_{\Theta} S_W}{|F|^2 S_{\Theta} + S_W} d\mathbf{w} \end{aligned}$$

Discussion on Deblurring

- Advantage of Wiener approach:
 - LTI filtering implementation
- Disadvantages of Wiener approach:
 - No natural way to incorporate nonnegativity constraints (in image processing, for instance)
 - Only truly optimal for Gaussian statistics
- Competing approaches include iterative methods such as the “Richardson-Lucy” algorithm (an EM-style procedure)
 - Computationally intensive
 - Can naturally incorporate nonnegativity
 - Sometimes better match to real statistics

Real-Time Wiener Filtering

- What if we don't have “future” measurements?
 - Must restrict h to be causal

- Solution:

$$H(z) = \frac{1}{S_Y^-(z)} \left\{ \frac{S_{\Theta Y}(z)}{S_Y^+(z)} \right\}_+$$

where the meaning of the plus and minus superscripts and subscripts will be defined on later slides

Spectral Factorization

- If Y has a spectrum satisfying the Paley-Wiener criterion:

$$\int_{-p}^p \log S_Y(\mathbf{w}) d\mathbf{w} > -\infty$$

then the spectrum can be factored as

$$S_Y(\mathbf{w}) = S_Y^+(\mathbf{w}) S_Y^-(\mathbf{w})$$

where

$$\mathcal{F}^{-1}\{S_Y^+\} \text{ is causal}$$

$$\mathcal{F}^{-1}\{S_Y^-\} \text{ is anticausal}$$

Factoring Rational Spectra

- If the spectrum is a ratio of polynomials, we can factor as

$$S_Y(z) = \underbrace{S_Y^+(z)}_{\substack{\text{Poles and zeros} \\ \text{inside unit circle}}} \underbrace{S_Y^-(z)}_{\substack{\text{Poles and zeros} \\ \text{outside unit circle}}} = S_Y^+(z) S_Y^+(z^{-1})$$

- Aside: spectral factorization into causal and anticausal factors is analogous to Cholesky decomposition of a covariance matrix into lower and upper triangular factors

Causal Part Extraction

- We can split f into its causal and anticausal parts:

$$f(k) = \underbrace{\{f(k)\}_+}_{\text{causal}} + \underbrace{\{f(k)\}_-}_{\text{anticausal}}$$

$$f(k)_+ = f(k)u(k), \quad \{f(k)\}_- = f(k)u(-k-1)$$

- Use similar notation for Z-transform domain

$$F(z) = \{F(z)\}_+ + \{F(z)\}_-$$

$$\{F\}_+ = \mathcal{Z} \{ \mathcal{Z}^{-1} \{F\} u(k) \}$$

$$\{F\}_- = \mathcal{Z} \{ \mathcal{Z}^{-1} \{F\} u(-k-1) \}$$

How to Extract Causal Parts

- If F is a ratio of polynomials can usually do a partial fraction expansion:

$$F(z) = \underbrace{\{F(z)\}_+}_{\substack{\text{Poles and zeros} \\ \text{inside unit circle}}} + \underbrace{\{F(z)\}_-}_{\substack{\text{Poles and zeros} \\ \text{outside unit circle}}}$$

- Can also do polynomial long division (see Ed Kamen's book)
- Almost always a total pain and really annoying